Boost 1channel white LED driver
For large LCDs

BD9285F

General Description
BD9285F is a high efficiency driver for white LEDs and designed for large LCDs. This IC is built-in a boost DCDC converters that employ an array of LEDs as the light source. BD9285F has some protect function against fault conditions, such as the over-voltage protection (OVP), the over current limit protection of DCDC (OCP), LED over current protection (LEDOCP), the open detection of LED string. Therefore BD9285F is available for the fail-safe design over a wide range output voltage.

Features
- Current mode DCDC converter
- Vout discharge circuit as shutdown
- LED protection circuit (OPEN protection, LED OCP protection)
- LED protect detection as small PWM dimming signal
- Over-voltage protection (OVP) and low-voltage protection (SCP: short circuit protection) for the output voltage Vout
- Adjustable soft start time constant
- The wide range of analog dimming 0.2V-3.5V
- The built-in transformation circuit from pulse to DC
- 2 PWM dimming signal
- The UVLO detection for the input voltage of the power stage
- FAIL logic output

Applications
- TV, PC display and other LCD backlight system.

Key Specifications
- Input voltage range: 9.0V to 18.0V
- DCDC oscillation frequency: 150kHz (RT=100kΩ)
- Active current consumption: 1.2mA(Typ.)
- Operating temperature range: -40°C to +85°C

Package(s)
- SOP18
  - W(Typ.) x D(Typ.) x H(Max.)
  - 11.20mm x 7.80mm x 2.01mm
  - Pin pitch 1.27mm

Figure 1. SOP18

Typical Application Circuit(s)

Figure 2. Typical application circuit
Absolute Maximum Ratings (Ta=25℃)

<table>
<thead>
<tr>
<th>Parameter</th>
<th>Symbol</th>
<th>Ratings</th>
<th>Unit</th>
</tr>
</thead>
<tbody>
<tr>
<td>Input voltage</td>
<td>Vccmax</td>
<td>20</td>
<td>V</td>
</tr>
<tr>
<td>STB pin voltage</td>
<td>STB</td>
<td>20</td>
<td>V</td>
</tr>
<tr>
<td>OVP, UVLO, SS, RT, ISENSE, FB, CS, TC54 pin voltage</td>
<td>OVP, UVLO, SS, RT, ISENSE, FB, CS, TC54</td>
<td>7</td>
<td>V</td>
</tr>
<tr>
<td>PWM1, PWM2, FAILB, ADIM, ADIM., P pin voltage</td>
<td>PWM1, PWM2, FAILB, ADIM, ADIM., P</td>
<td>20</td>
<td>V</td>
</tr>
<tr>
<td>DIMOUT, GATE pin voltage</td>
<td>DIMOUT, GATE</td>
<td>Vcc</td>
<td>V</td>
</tr>
<tr>
<td>Power Dissipation</td>
<td>Pd</td>
<td>687 (*1)</td>
<td>mW</td>
</tr>
<tr>
<td>Operating Temperature Range</td>
<td>Topr</td>
<td>-40 to +85</td>
<td>℃</td>
</tr>
<tr>
<td>Junction Temperature</td>
<td>Tjmax</td>
<td>150</td>
<td>℃</td>
</tr>
<tr>
<td>Storage Temperature Range</td>
<td>Tstg</td>
<td>-55 to +150</td>
<td>℃</td>
</tr>
</tbody>
</table>

*1 Pd derated at 5.5 mW/℃ for temperature above Ta=25℃, mounted on 70mm × 70mm × 1.6mm 1 layer glass-epoxy PCB.

Operation range

<table>
<thead>
<tr>
<th>Parameter</th>
<th>Symbol</th>
<th>Range</th>
<th>Unit</th>
</tr>
</thead>
<tbody>
<tr>
<td>VCC Power source voltage</td>
<td>VCC</td>
<td>9.0 to 18.0</td>
<td>V</td>
</tr>
<tr>
<td>DC/DC oscillation frequency</td>
<td>fsw</td>
<td>50 to 800</td>
<td>kHz</td>
</tr>
<tr>
<td>The effective range of ADIM signal</td>
<td>VADIM</td>
<td>0.2 to 3.5</td>
<td>V</td>
</tr>
<tr>
<td>PWM input frequency range</td>
<td>FPWM</td>
<td>100 to 100k</td>
<td>Hz</td>
</tr>
</tbody>
</table>

Pin Configuration

Package dimension, marking diagram

Figure 3-1. Pin configuration

Figure 3-2. Package dimension
### 1.1 Electrical character (Unless otherwise specified Ta=25°C, VCC=12V)

<table>
<thead>
<tr>
<th>Parameter</th>
<th>Symbol</th>
<th>Limit</th>
<th>Unit</th>
<th>Condition</th>
</tr>
</thead>
<tbody>
<tr>
<td><strong>Total current consumption</strong></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>Circuit current</td>
<td>Icc</td>
<td>-</td>
<td>1.2</td>
<td>1.8 mA VSTB=3V, PWM1=PWM2=0V</td>
</tr>
<tr>
<td>Standby current</td>
<td>IST</td>
<td>-</td>
<td>0</td>
<td>3 μA VSTB=0V</td>
</tr>
<tr>
<td><strong>UVLO block</strong></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>Operation voltage (VCC)</td>
<td>VUVLO_VCC</td>
<td>6.5</td>
<td>7.5</td>
<td>8.5 V VCC=Sweep UP</td>
</tr>
<tr>
<td>Hysteresis Voltage (VCC)</td>
<td>VUHYS_VCC</td>
<td>150</td>
<td>300</td>
<td>600 mV VCC=Sweep DOWN</td>
</tr>
<tr>
<td>UVLO release voltage</td>
<td>VUVLO</td>
<td>2.88</td>
<td>3.00</td>
<td>3.12 V VUVLO=Sweep UP</td>
</tr>
<tr>
<td>UVLO hysteresis voltage</td>
<td>VUHYS</td>
<td>160</td>
<td>200</td>
<td>240 mV VUVLO=Sweep DOWN</td>
</tr>
<tr>
<td>UVLO pin leak current</td>
<td>UVLO_LK</td>
<td>-2</td>
<td>0</td>
<td>2 μA VUVLO=4V</td>
</tr>
<tr>
<td><strong>DC/DC block</strong></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>ISENSE threshold voltage 1</td>
<td>VLED1</td>
<td>1.47</td>
<td>1.50</td>
<td>1.53 V VADIM=1.5V</td>
</tr>
<tr>
<td>ISENSE threshold voltage 2</td>
<td>VLED2</td>
<td>3.33</td>
<td>3.50</td>
<td>3.67 V VADIM=5.0V (as mask analog dimming)</td>
</tr>
<tr>
<td>ISENSE threshold voltage 3</td>
<td>VLED3</td>
<td>-2</td>
<td>-</td>
<td>+2 % VADIM=0.7V</td>
</tr>
<tr>
<td>Oscillation frequency</td>
<td>FCT</td>
<td>142.5</td>
<td>150</td>
<td>157.5 kHz RT=100kohm</td>
</tr>
<tr>
<td>GATE pin MAX DUTY output</td>
<td>NMAX_DUTY</td>
<td>90</td>
<td>95</td>
<td>99 % RT=100kohm</td>
</tr>
<tr>
<td>GATE pin ON resistance (as source)</td>
<td>RONSO</td>
<td>3.0</td>
<td>6.0</td>
<td>12.0 Ω ION=-10mA</td>
</tr>
<tr>
<td>GATE pin ON resistance (as sink)</td>
<td>RONSI</td>
<td>1.2</td>
<td>2.5</td>
<td>5.0 Ω ION=10mA</td>
</tr>
<tr>
<td>RT pin voltage</td>
<td>VRT</td>
<td>1.0</td>
<td>1.5</td>
<td>2.0 V RT=100kohm</td>
</tr>
<tr>
<td>SS pin source current</td>
<td>ISSSO</td>
<td>-4.20</td>
<td>-3.0</td>
<td>-2.14 μA VSS=2V</td>
</tr>
<tr>
<td>SS pin Low output voltage</td>
<td>VSS_L</td>
<td>-</td>
<td>0.20</td>
<td>0.50 V VSTB=0V, Ioss=50μA</td>
</tr>
<tr>
<td>Soft start ended voltage</td>
<td>VSS_END</td>
<td>2.7</td>
<td>3.0</td>
<td>3.3 V SS=Sweep UP</td>
</tr>
<tr>
<td>FB source current</td>
<td>IFBSO</td>
<td>-140</td>
<td>-100</td>
<td>-60 μA VISENSE=0.2V, VFB=1.0V</td>
</tr>
<tr>
<td>FB sink current</td>
<td>IFBSI</td>
<td>60</td>
<td>100</td>
<td>140 μA VISENSE=2.0V, VFB=1.0V</td>
</tr>
<tr>
<td>OCP detect voltage</td>
<td>VCS</td>
<td>450</td>
<td>500</td>
<td>550 mV CS=Sweep UP</td>
</tr>
<tr>
<td><strong>DC/DC protection block</strong></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>OVP detect voltage</td>
<td>VOVP</td>
<td>2.88</td>
<td>3.00</td>
<td>3.32 V VOVP Slewweep UP</td>
</tr>
<tr>
<td>OVP detect hysteresis</td>
<td>VOVP_HYS</td>
<td>50</td>
<td>100</td>
<td>150 mV VOVP Slewweep DOWN</td>
</tr>
<tr>
<td>SCP detect voltage</td>
<td>VSCP</td>
<td>0.14</td>
<td>0.20</td>
<td>0.26 V VOVP Slewweep DOWN</td>
</tr>
<tr>
<td>SCP detect hysteresis</td>
<td>VSCP_HYS</td>
<td>25</td>
<td>60</td>
<td>75 mV VOVP Slewweep DOWN</td>
</tr>
<tr>
<td>OVP pin leak current</td>
<td>OVP_LK</td>
<td>-2</td>
<td>0</td>
<td>2 μA VOVP=4V</td>
</tr>
</tbody>
</table>
## 1.2 Electrical character (Unless otherwise specified T_a=25°C, VCC=12V)

<table>
<thead>
<tr>
<th>Parameter</th>
<th>Symbol</th>
<th>Limit</th>
<th>Unit</th>
<th>Condition</th>
</tr>
</thead>
<tbody>
<tr>
<td><strong>[LED protection block]</strong></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>LED OCP detect voltage</td>
<td>VLEDSCP</td>
<td>3.8</td>
<td>4.0</td>
<td>4.2</td>
</tr>
<tr>
<td>LED OPEN detect voltage</td>
<td>VOPEN</td>
<td>0.05</td>
<td>0.10</td>
<td>0.15</td>
</tr>
<tr>
<td><strong>[Analog dimming block]</strong></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>ADIM_P pin HIGH voltage</td>
<td>ADIM_PH</td>
<td>2.0</td>
<td>-</td>
<td>3.8</td>
</tr>
<tr>
<td>ADIM_P pin LOW voltage</td>
<td>ADIM_PL</td>
<td>-0.3</td>
<td>-</td>
<td>0.8</td>
</tr>
<tr>
<td>ADIM_P pin input mask voltage</td>
<td>ADIM_PPU</td>
<td>4.2</td>
<td>-</td>
<td>5.6</td>
</tr>
<tr>
<td>ADIM_P pin pull-down resistance</td>
<td>RADIM_P</td>
<td>130</td>
<td>200</td>
<td>300</td>
</tr>
<tr>
<td>ADIM pin output voltage H</td>
<td>ADIMH</td>
<td>3.201</td>
<td>3.30</td>
<td>3.399</td>
</tr>
<tr>
<td>ADIM pin output voltage L</td>
<td>ADIML</td>
<td>-0.0</td>
<td>0.05</td>
<td>V</td>
</tr>
<tr>
<td>ADIM pin output resistance</td>
<td>ADIMR</td>
<td>6.6</td>
<td>10</td>
<td>15</td>
</tr>
<tr>
<td>ADIM pin leak current</td>
<td>ILADIM</td>
<td>-2</td>
<td>0</td>
<td>2</td>
</tr>
<tr>
<td>ISENSE pin leak current</td>
<td>IL_ISENSE</td>
<td>-2</td>
<td>0</td>
<td>2</td>
</tr>
<tr>
<td><strong>[Dimming signal output block]</strong></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>DIMOUT source on-resistance</td>
<td>RONSO</td>
<td>6.0</td>
<td>12.0</td>
<td>24.0</td>
</tr>
<tr>
<td>DIMOUT sink on-resistance</td>
<td>RONSI</td>
<td>1.7</td>
<td>3.5</td>
<td>7.0</td>
</tr>
<tr>
<td><strong>[TC54 block]</strong></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>TC54 output voltage</td>
<td>VTC54</td>
<td>5.2</td>
<td>5.4</td>
<td>5.6</td>
</tr>
<tr>
<td>TC54 available current</td>
<td>ITC54</td>
<td>100</td>
<td>-</td>
<td>-</td>
</tr>
<tr>
<td>TC54_UVLO detect voltage</td>
<td>TC54_TH</td>
<td>2.232</td>
<td>2.4</td>
<td>2.568</td>
</tr>
<tr>
<td>TC54_UVLO hysteresis</td>
<td>TC54_HYS</td>
<td>50</td>
<td>100</td>
<td>200</td>
</tr>
<tr>
<td>TC54 discharge current</td>
<td>TC54_DIS</td>
<td>5</td>
<td>10</td>
<td>15</td>
</tr>
<tr>
<td><strong>[STB block]</strong></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>STB pin HIGH voltage</td>
<td>STBH</td>
<td>-2.2</td>
<td>-</td>
<td>19</td>
</tr>
<tr>
<td>STB pin LOW voltage</td>
<td>STBL</td>
<td>-0.3</td>
<td>-</td>
<td>0.8</td>
</tr>
<tr>
<td>STB pin input current</td>
<td>ISTB</td>
<td>2.0</td>
<td>3.0</td>
<td>4.5</td>
</tr>
<tr>
<td><strong>[PWM block]</strong></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>PWM pin HIGH Voltage</td>
<td>PWMM</td>
<td>2.0</td>
<td>-</td>
<td>5.0</td>
</tr>
<tr>
<td>PWM pin LOW Voltage</td>
<td>PWM_L</td>
<td>-0.3</td>
<td>-</td>
<td>0.8</td>
</tr>
<tr>
<td>PWM x pin Pull-Down resistance</td>
<td>RPWM</td>
<td>130</td>
<td>200</td>
<td>300</td>
</tr>
<tr>
<td><strong>[FAIL block (OPEN DRAIN)]</strong></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>FAILB pin on-resistance</td>
<td>RFAIL</td>
<td>0.75</td>
<td>1.5</td>
<td>3.0</td>
</tr>
<tr>
<td>FAILB pin leak current</td>
<td>ILFAIL</td>
<td>-2</td>
<td>0</td>
<td>2</td>
</tr>
</tbody>
</table>
### 1.3 Pin number, pin name, pin function

<table>
<thead>
<tr>
<th>No.</th>
<th>name</th>
<th>IN/OUT</th>
<th>function</th>
<th>rating [V]</th>
</tr>
</thead>
<tbody>
<tr>
<td>1</td>
<td>OVP</td>
<td>In</td>
<td>Over voltage protection detection pin</td>
<td>-0.3 to 7</td>
</tr>
<tr>
<td>2</td>
<td>UVLO</td>
<td>In</td>
<td>Under voltage lock out detection pin</td>
<td>-0.3 to 7</td>
</tr>
<tr>
<td>3</td>
<td>SS</td>
<td>Out</td>
<td>Slow start setting pin</td>
<td>-0.3 to 7</td>
</tr>
<tr>
<td>4</td>
<td>RT</td>
<td>Out</td>
<td>For DC/DC switching frequency setting pin</td>
<td>-0.3 to 7</td>
</tr>
<tr>
<td>5</td>
<td>PWM1</td>
<td>In</td>
<td>External PWM dimming signal input pin1</td>
<td>-0.3 to 20</td>
</tr>
<tr>
<td>6</td>
<td>PWM2</td>
<td>In</td>
<td>External PWM dimming signal input pin2</td>
<td>-0.3 to 20</td>
</tr>
<tr>
<td>7</td>
<td>FAILB</td>
<td>Out</td>
<td>Abnormality detection output pin</td>
<td>-0.3 to 20</td>
</tr>
<tr>
<td>8</td>
<td>ADIM</td>
<td>In/Out</td>
<td>ADIM signal input-output pin</td>
<td>-0.3 to 20</td>
</tr>
<tr>
<td>9</td>
<td>ADIM_P</td>
<td>In</td>
<td>ADIM pulse signal input pin</td>
<td>-0.3 to 20</td>
</tr>
<tr>
<td>10</td>
<td>GND</td>
<td>-</td>
<td>-</td>
<td>-</td>
</tr>
<tr>
<td>11</td>
<td>DIMOUT</td>
<td>Out</td>
<td>Dimming signal pin for driving MOSFET</td>
<td>-0.3 to VCC</td>
</tr>
<tr>
<td>12</td>
<td>GATE</td>
<td>Out</td>
<td>DC/DC switching output pin</td>
<td>-0.3 to VCC</td>
</tr>
<tr>
<td>13</td>
<td>STB</td>
<td>In</td>
<td>IC On/OFF pin</td>
<td>-0.3 to 20</td>
</tr>
<tr>
<td>14</td>
<td>VCC</td>
<td>-</td>
<td>Power supply pin</td>
<td>-0.3 to 20</td>
</tr>
<tr>
<td>15</td>
<td>ISENSE</td>
<td>In</td>
<td>Current detection input pin</td>
<td>-0.3 to 7</td>
</tr>
<tr>
<td>16</td>
<td>FB</td>
<td>In/Out</td>
<td>Error amplifier output pin</td>
<td>-0.3 to 7</td>
</tr>
<tr>
<td>17</td>
<td>CS</td>
<td>In</td>
<td>DC/DC output current detect pin, OCP input pin</td>
<td>-0.3 to 7</td>
</tr>
<tr>
<td>18</td>
<td>TC54</td>
<td>Out</td>
<td>5.4V output pin, shutdown timer pin</td>
<td>-0.3 to 7</td>
</tr>
</tbody>
</table>
2.1.1 Pin ESD Type

<table>
<thead>
<tr>
<th>OVP</th>
<th>UVLO</th>
<th>SS</th>
</tr>
</thead>
<tbody>
<tr>
<td><img src="image1.png" alt="OVP Circuit" /></td>
<td><img src="image2.png" alt="UVLO Circuit" /></td>
<td><img src="image3.png" alt="SS Circuit" /></td>
</tr>
</tbody>
</table>

<table>
<thead>
<tr>
<th>RT</th>
<th>PWM1, PWM2</th>
<th>FAILB</th>
</tr>
</thead>
<tbody>
<tr>
<td><img src="image4.png" alt="RT Circuit" /></td>
<td><img src="image5.png" alt="PWM1, PWM2 Circuit" /></td>
<td><img src="image6.png" alt="FAILB Circuit" /></td>
</tr>
</tbody>
</table>

<table>
<thead>
<tr>
<th>ADIM</th>
<th>ADIM_P</th>
<th>DIMOUT</th>
</tr>
</thead>
<tbody>
<tr>
<td><img src="image7.png" alt="ADIM Circuit" /></td>
<td><img src="image8.png" alt="ADIM_P Circuit" /></td>
<td><img src="image9.png" alt="DIMOUT Circuit" /></td>
</tr>
</tbody>
</table>

<table>
<thead>
<tr>
<th>GATE</th>
<th>STB</th>
<th>ISENSE</th>
</tr>
</thead>
<tbody>
<tr>
<td><img src="image10.png" alt="GATE Circuit" /></td>
<td><img src="image11.png" alt="STB Circuit" /></td>
<td><img src="image12.png" alt="ISENSE Circuit" /></td>
</tr>
</tbody>
</table>

Figure 4-1. Internal equivalent circuit
2.1.2 Pin ESD Type

<table>
<thead>
<tr>
<th>FB</th>
<th>CS</th>
<th>TC54</th>
</tr>
</thead>
<tbody>
<tr>
<td><img src="internal_equivalent_circuit.png" alt="" /></td>
<td><img src="internal_equivalent_circuit.png" alt="" /></td>
<td><img src="internal_equivalent_circuit.png" alt="" /></td>
</tr>
</tbody>
</table>

Figure 4-2. Internal equivalent circuit

2.2 Block diagram

Figure 5. Block diagram
2.3 Typical performance Curves

![Graph 1: Operating current (ICC) vs VCC](Figure 6. Operating current (ICC) vs VCC)

![Graph 2: GATE frequency vs RT](Figure 7. GATE frequency vs RT)

![Graph 3: FB sink current vs FB voltage](Figure 8. FB sink current vs FB voltage)

![Graph 4: FB source current vs FB voltage](Figure 9. FB source current vs FB voltage)

![Graph 5: ISENSE feedback voltage vs ADIM](Figure 10. ISENSE feedback voltage vs ADIM)
2.4 Pin function description

OPin1: OVP
The OVP terminal is the input for over-voltage protection and short circuit protection of output voltage. As OVP is more than 3.0V, the over-voltage protection (OVP) will work. On the other hand, OVP is lower than 0.2V, the short circuit protection (SCP) will work. At the moment of these detections, the BD9285 stops the switching of the output GATE and starts to count up the abnormal interval, but IC doesn't reach latch off state instantaneously until the detection continues up to the number of counts of GATE terminals, which depend on the kind of abnormality. (Please refer to the time chart in the section 3.5.7)

The OVP pin is high impedance, because the internal resistance to a certain bias is not connected.
So, the bias by the external components is required, even if OVP function is not used, because the open connection of this pin is not fixed the potential.
The setting examples is separately described in the section 3.4.6, "external components selection, how to set OVP, SCP"

OPin2: UVLO
Under voltage lock out pin for the input voltage of the power stage. More than 3.0V(typ.), IC starts the boost operation and stops lower than 2.8V(typ.).
The UVLO pin is high impedance, because the internal resistance to a certain bias is not connected.
So, the bias by the external components is required, even if UVLO function is not used, because the open connection of this pin is not fixed the potential.
The setting examples is separately described in the section 3.4.5, "external components selection, how to set UVLO"

OPin3: SS
The pin which sets soft start interval of DC/DC converter. It performs the constant current charge of 3.0 μA to external capacitance Css(0.001μF to 4.7μF). The switching duty of GATE output will be limited during 0V to 3.0V of the SS voltage.
So the equality of the soft start interval can be expressed as following
\[ T_{ss} = 1.0 \times 10^6 \times C_{ss} \]
Css: the external capacitance of the SS pin.
Regarding of the logic of SS=L
\[(SS=L) = (PWM1 and PWM2 have not asserted H since ResetB=L->H) or (latch off state)\]
where ResetB = (STB=H) and (VCCUVLO=H) and (UVLO=H) and (TC54UVLO=H)
Please refer to the time chart on soft start behavior in the section 3.7.4

OPin4: RT
DC/DC switching frequency setting pin. RT set the oscillation frequency inside IC.
The relationship between the frequency and RT resistance value (ideal)
\[ R_{RT} = \frac{15000}{f_{SW} \ [kHz]} \ [k\Omega] \]
The oscillation setting range from 50kHz to 800kHz.
The setting examples is separately described in the section 3.4.4, "external components selection, how to set DCDC oscillation frequency"

OPin5, Pin6: PWM1, PWM2
The ON / OFF terminal of the LED driver. LED lights when both PWM signal are high (DIMOUT = H). The Duty signal of this pin can control the PWM dimming.
The high / low level of PWM pins are following.

<table>
<thead>
<tr>
<th>State</th>
<th>PWM input voltage</th>
</tr>
</thead>
<tbody>
<tr>
<td>PWM1=H or PWM2=H</td>
<td>PWM=2.0V to 5.0V</td>
</tr>
<tr>
<td>PWM1=L or PWM2=L</td>
<td>PWM=-0.3V to 0.8V</td>
</tr>
</tbody>
</table>
PWM1 and PWM2 have the functional difference, and GATE pin outputs only by the logic of PWM1. This is why only boost operation continues while PWM1=H, PWM2=L. In this case, the adequate confirmation is required not to be over voltage of the output voltage Vout.

Figure 11. PWM pin function

Pin7: FAILB
FAIL signal output pin (open drain). As abnormal, the internal NMOS turn on.

<table>
<thead>
<tr>
<th>Status</th>
<th>FAILB output</th>
</tr>
</thead>
<tbody>
<tr>
<td>Normal</td>
<td>OPEN</td>
</tr>
<tr>
<td>Abnormal</td>
<td>GND Level</td>
</tr>
</tbody>
</table>

Pin8: ADIM
The input output pin for analog dimming signal. The pin function can be changed according to the input level of ADIM_P pin. The pulse-DC transform circuit is included into BD9285F.

<table>
<thead>
<tr>
<th>ADIM_P input level</th>
<th>ADIM_P pin function</th>
<th>ADIM pin function</th>
<th>Required signal to IC</th>
</tr>
</thead>
<tbody>
<tr>
<td>-0.3V&lt;ADIM_P&lt;3.8V</td>
<td>Pulse signal input for analog dimming</td>
<td>DC output signal for analog dimming</td>
<td>DUTY signal for analog dimming</td>
</tr>
<tr>
<td>4.2V&lt;ADIM_P&lt;5.6V</td>
<td>ADIM_P pin function is masked.</td>
<td>DC input signal for analog dimming</td>
<td>DC signal for analog dimming</td>
</tr>
</tbody>
</table>

Figure 12. Analog dimming function and character

Above functions enable BD9285 use both of the duty and DC signal for analog dimming.

When the duty signal is used, that input to the pin ADIM_P with the amplitude about 3.3V. The input duty of ADIM_P needs to be larger than 6% so that the output ADIM is larger than 0.2V. In the case of the normal feedback with analog dimming, The ADIM pin voltage is equal to the ISENSE pin voltage. Therefore, please be careful that the lower ADIM voltage than 0.1V causes the OPEN abnormal detection.

When the DC signal is used, ADIM_P will be pulled up, and the signal input to the pin ADIM.
In the driver module with more than two BD9285, and the analog dimming is performed by the duty signal, the architecture will be shown in the right figure. That can reduce the LED current error between the channels, because the common circuit of the pulse DC transform is used.

The pulse DC transform circuit outputs DC signal to the ADIM pin with the time constant of R1, C1 in the above diagram. More C1 value, the ripple components of the ADIM pin

Figure 13. the analog dimming circuit as two BD9285 are used.
is decreased, on the other hand, the transient response is delayed. And please keep in mind the error voltage if the pull down resistor of ADIM pin will be connected.

**Pin9: ADIM_P**
The pulse signal input pin for analog dimming. Please pull up the voltage level more than 4.2V(typ.), when DC signal is used for the analog dimming. In normal operation, please set the input voltage under 5.6V. For more details, please refer to <ADIM> pin descriptions. The input frequency of this pin assumed from 2kHz to 100kHz. Please keep in mind that the capacitor of ADIM pin is small considering of this input frequency, the error of LED current can be cause.

**Pin10: GND**
GND pin of IC.

**Pin11: DIMOUT**
This is the output pin for external NMOS of dimming. The below table shows the rough output logic of each operation state, and the output H level is VCC. Please refer to the time chart in the section 3.7 for detail explanations, because The DIMOUT logic has the exceptional behavior. Please insert the resistance between the dimming MOS gate to improve the overshoot of LED current, as PWM turns from low to high.

<table>
<thead>
<tr>
<th>Status</th>
<th>DIMOUT output</th>
</tr>
</thead>
<tbody>
<tr>
<td>Normal</td>
<td>PWM1 and PWM2</td>
</tr>
<tr>
<td>Abnormal</td>
<td>GND Level</td>
</tr>
</tbody>
</table>

**Pin12: GATE**
This is the output terminal for driving the gate of the boost MOSFET. The high level is VCC of IC. Frequency can be set by the resistor connected to RT. Please refer to the <RT> pin description for the frequency setting.

**Pin13: STB**
ON/OFF setting terminal for IC, which can be used to perform a reset at shutdown. Please reset this pin after latch off. Regarding of the sequence of turning on, if the input logic STB turns from low to high, the internal power supply is activated. After the positive edge of PWM is input, BD9285 starts the boost operation.

- The input voltage of STB pin toggles the IC state(IC ON/OFF). Please avoid the use of the intermediate level (from 0.8V to 2.2V).

Regarding of the power down sequence, while STB=L and TC54UVLO=H, in order to discharge the output voltage, DIMOUT logic can assert high, depending on the PWM logic. This discharge behavior is separately described in the time chart in the section 3.7.3, or in the section 3.4.2, "how to shutdown and set TC54 capacitance"

**Pin14: VCC**
Power supply pin of IC. Input range is from 9V to 18.0V. The operation starts more than 7.5 V(TYP.) and shuts down less than 7.2 V(TYP.).

**Pin15: ISENSE**
This is the input terminal for the current detection. The error amplifier compares the lower voltage the analog dimming pin ADIM and 3.5V. The abnormal voltage of this pin activates the protection function of LED, such as LEDOCP, OPEN.

**[LED OCP Protection Function]**
More than ISENSE = 4.0V (typ.), the over current of LED (LEDOCP) will be detected. If that states continues 4096 clock of GATE pin, IC will latch off. (Please refer to the time chart in the section 3.7.8.)

**[LED OPEN Protection Function]**
If OPEN state (ISENSE<0.1V) continues during 4 clocks interval of GATE terminal, BD9285 starts to count the interval of the abnormal state. In that counting state, DIMOUT logic keeps high output no matter what PWM logic so that the OPEN abnormal state can be detected continuously. If the abnormal condition continues by the completion of counting, BD9285 will be latched off. (Please refer to the time chart in the section 3.7.7.)

Exceptionally the OPEN protect detection are masked in the following conditions, CASE1. When PWM = L. ISENSE is less than 0.1V even in normally, because DIMOUT = L.
CASE2. In the soft-start interval. ISENSE is less than 0.1V, because of the insufficient output voltage Vout.

**Pin16: FB**
This is the output terminal of error amplifier. Monitoring the ISENSE terminal voltage, this pin outputs the error signal with the analog dimming signal (pin ADIM) or 3.5V. After the completion of the SS, this pin outputs high impedance as the logic “PWM1 and PWM2” asserts low. FB voltage

---

**Figure 14. ISENSE pin circuit**
is hold to the external capacitance.
(For more detail on the compensation setting is described in the section "3.6 loop compensation").

**Pin17: CS**
The CS pin has two functions.

1. **DC / DC current mode Feedback terminal**
The inductor current is converted to the CS pin voltage by the sense resistor $R_{CS}$ and this CS pin voltage controls the output voltage by compared with the error amp output.

2. **Inductor current limit (OCP) terminal**
The CS terminal also has a over current protection (OCP), if it voltage is more than 0.5V, the switching operation will be stopped compulsorily.

   Both of above functions are enable after 300ns (typ.) when GATE pin asserts high, because the leading Edge Blanking function is included into this IC to prevent the affect noise. Please refer to the section 3.5.1 “DCDC parts selection / how to set OCP”, for detail explanation.

**Pin18: TC54**
This is the 5.4V (TYP.) output pin that is used for internal power supply. Available current is 100uA

TC54 can be used as a timer for the discharge of output capacitance DCDC. For detailed instructions, please refer the section 3.4.2 “how to shutdown and set TC54 capacitance”
3.1 Application circuit example
The bellows are example circuits of using BD9285F.

- The basic application circuit example

![Basic Application Circuit Example](Figure 16)

- As for the dimming signal, the single PWM and the DC for analog dimming

![Circuit Example with Single PWM](Figure 17)

![Circuit Example with Single PWM](Figure 18)

PWM1 pin and PWM2 pin are shorted.
- Only analog dimming

Figure 19. the circuit example of analog dimming only

- Application example when use numerous IC
The application circuit of analog dimming by external duty signal.

Figure 20. the circuit example of when plural IC is used.
### 3.2 The detection condition list of the protection (TYP. Condition)

<table>
<thead>
<tr>
<th>Protection</th>
<th>Detect pin condition</th>
<th>Detect condition</th>
<th>Release condition</th>
<th>Timer operation</th>
<th>Protection type</th>
</tr>
</thead>
<tbody>
<tr>
<td>LED OPEN</td>
<td>ISENSE &lt; 0.1V</td>
<td>PWM1 and PWM2</td>
<td>SS &gt; 3.0V</td>
<td>4096 count</td>
<td>Latch off</td>
</tr>
<tr>
<td>LED OCP</td>
<td>ISENSE &gt; 4.0V</td>
<td>-</td>
<td>ISENSE &lt; 4.0V</td>
<td>4096 count</td>
<td>Latch off</td>
</tr>
<tr>
<td>UVLO</td>
<td>UVLO &lt; 2.8V</td>
<td>-</td>
<td>UVLO &gt; 3.0V</td>
<td>NO</td>
<td>Auto recovery</td>
</tr>
<tr>
<td>TC54 UVLO</td>
<td>TC54 &lt; 2.4V</td>
<td>-</td>
<td>TC54 &gt; 2.5V</td>
<td>NO</td>
<td>Auto recovery</td>
</tr>
<tr>
<td>VCC UVLO</td>
<td>VCC &lt; 7.2V</td>
<td>-</td>
<td>VCC &gt; 7.5V</td>
<td>NO</td>
<td>Auto recovery</td>
</tr>
<tr>
<td>OVP</td>
<td>OVP &gt; 3.0V</td>
<td>-</td>
<td>OVP &lt; 2.9V</td>
<td>4 count</td>
<td>Latch off</td>
</tr>
<tr>
<td>SCP</td>
<td>OVP &lt; 0.2V</td>
<td>SS &gt; 3.0V</td>
<td>OVP &gt; 0.25V</td>
<td>4096 count</td>
<td>Latch off</td>
</tr>
<tr>
<td>OCP</td>
<td>CS &gt; 0.5V</td>
<td>-</td>
<td>-</td>
<td>NO</td>
<td>Pulse by Pulse</td>
</tr>
</tbody>
</table>

To reset the latch type protection, please input of STB logic to ‘L’ once. Otherwise the detection of VCCUVLO, TC54UVLO is required.

### 3.3 The behavior list of the protection

<table>
<thead>
<tr>
<th>Protect Function</th>
<th>DC/DC Gate output</th>
<th>Dimming transistor (DIMOUT) logic</th>
<th>Soft Start</th>
<th>FAILB pin</th>
</tr>
</thead>
<tbody>
<tr>
<td>LED OPEN</td>
<td>Stops after latch</td>
<td>H after 4clk, L after latch</td>
<td>discharge after latch</td>
<td>L after latch</td>
</tr>
<tr>
<td>LED OCP</td>
<td>Stops immediately</td>
<td>H immediately, L after latch</td>
<td>discharge after latch</td>
<td>L after latch</td>
</tr>
<tr>
<td>STB</td>
<td>Stops immediately</td>
<td>L if TC54 &lt; 2.4V</td>
<td>discharge immediately</td>
<td>OPEN</td>
</tr>
<tr>
<td>UVLO</td>
<td>Stops immediately</td>
<td>immediately L</td>
<td>discharge immediately</td>
<td>immediately L</td>
</tr>
<tr>
<td>TC54 UVLO</td>
<td>Stops immediately</td>
<td>immediately L</td>
<td>discharge immediately</td>
<td>immediately L</td>
</tr>
<tr>
<td>VCC UVLO</td>
<td>Stops immediately</td>
<td>immediately L</td>
<td>discharge immediately</td>
<td>immediately L</td>
</tr>
<tr>
<td>OVP</td>
<td>Stops immediately</td>
<td>immediately L</td>
<td>discharge after latch</td>
<td>L after latch</td>
</tr>
<tr>
<td>SCP</td>
<td>Stops immediately</td>
<td>immediately L</td>
<td>discharge after latch</td>
<td>L after latch</td>
</tr>
<tr>
<td>OCP</td>
<td>Stops immediately</td>
<td>Normal operation</td>
<td>Not discharge</td>
<td>OPEN</td>
</tr>
</tbody>
</table>

Please refer to the timing chart in the section 3.7 for the detail.

Not Recommended for New Designs
● 3.4 External components selection

● 3.4.1 The start up operation and the setting of Soft Start external capacitance

The below explanations are the start up sequency of BD9285.

![Figure 21. the turn-on waveform](image1)

![Figure 22. the turn-on circuit](image2)

○ The explanation of start up sequency

① When STB is H, the internal bias voltage of TC54 rising.
② With the first PWM=H, BD9285 enables output the boost pulse, and the SS start to charge to the external capacitance. At this moment, the voltage of FB will be the same as SS voltage internally regardless of the PWM logic.
③ The FB=SS voltage reach the bottom voltage of saw-toothed wave and the DC/DC start to output the pulse signal. Therefore the boost of VOUT is started.
④ VOUT is boosted to fixed level, and the LED current is rising.
⑤ When the LED current reached to fixed level, FB is removed from SS internally. The start up operation completed.
⑥ IC start the normal operation by sensing the voltage of ISENSE pin. When SS is more than 3.0V, even if the LED current does not flows, the clamped circuit of SS and FB is off, and the protect detection of SCP and OPEN starts.

○ The setting method of SS external capacitance

As above described, DC/DC stops when the PWM1=L. It means the boost operation only enabled within PWM1=H duration and SS time will be extented while boost with small PWM duty. Also the SS time is affected by the output capacitance, the LED current and application conditions.

T_{ss} is defined as the time for the SS voltage to reach to the FB feedback voltage. Please set the T_{ss} longer than Trise_min, which is the start up time of the minimum PWM duty.

When the FB voltage during LED turns on is expressed VFB, the equality on T_{ss} is the following.

\[
T_{ss} = \frac{C_{ss} [F] \times VFB [V]}{3 [\mu A]} [Sec]
\]

So please set the external capacitance to meet the T_{ss}>>Trise_min.
3.4.2 how to shutdown and set TC54 capacitance
This IC is equipped the discharge function when shutdown is operated.

Explanation of shutdown sequence
1. When STB=L, DC/DC and TC54 are stop.
2. When STB=L, TC54UVLO=H, the DIMOUT logic asserts the PWM logic. The voltage of TC54 (5.4V) will decrease by the constant current -10uA and is discharged to 2.4V.
3. VOUT will be discharged and ILED decresing.
4. When the voltage of TC54 pin is under 2.4V(typ.), the IC will shutdown.

The setting method of TC54 external capacitance
Please use below formula to calculate the shutdown time TOFF.

\[ T_{OFF} = \frac{C_{REG} \times 3.0 \text{[V]}}{10 \mu\text{A}} \] [Sec]

As shown the above, the PWM signal is required even after STB=L. The discharge interval of VOUT is the longest in the minimum PWM duty. Please set the Creg value with a enough timing margin from the end of the VOUT discharge to shutdown.

3.4.3 The LED current setting
LED current can be adjusted by setting the resistance \( R_{ISENSE} \) which connects to ISENSE pin.

The relationship between RISENSE and ILED current
With DC dimming
\[ R_{ISENSE} = \frac{ADIM[V]}{I_{LED}[A]} \Omega \]
Without DC dimming
\[ R_{ISENSE} = \frac{3.5[V]}{I_{LED}[A]} \Omega \]

[setting example]
If ILED current is 400mA as ADIM is 1.5V, we can calculate \( R_{ISENSE} \) as below.
\[ R_{ISENSE} = \frac{ADIM[V]}{I_{LED}[A]} = \frac{1.5[V]}{0.4[A]} = 3.75[\Omega] \]

Figure 23. the shutdown waveform and circuit

Figure 24. the example of LED current setting
3.4.4. how to set DCDC oscillation frequency

$R_{RT}$ which connects to RT pin set the oscillation frequency of DCDC.

- the relationship between OSC and $R_{RT}$ (ideal)

$$R_{RT} = \frac{15000}{f_{SW} \text{[kHz]}} \text{[kΩ]}$$

where $f_{SW}$ is the oscillation frequency of DCDC [kHz]

This equation is an ideal equation in which correction factors are not applied. The adequate verification with an actual set needs to be performed to set frequency precisely.

[setting example]
If DCDC oscillation frequency is 200kHz, we can calculate the $R_{RT}$ as below.

$$R_{RT} = \frac{15000}{200 \text{[kHz]}} = 15000 \text{[Ω]} = 75 \text{[kΩ]}$$

Figure 25. RT pin setting example
● 3.4.5. how to set UVLO
Under voltage lock out pin for the input voltage of the power stage. More than 3.0V(typ.), IC starts boost operation and stops lower than 2.8V(typ.).
The UVLO pin is high impedance, because the internal resistance to a certain bias is not connected.
So, the bias by the external components is required, even if UVLO function is not used, because the open connection of this pin is not fixed the potential.
The resistor value can be calculated by the below formula, if the VIN voltage is monitored, and that is divided by the resistor R1, R2 like the below diagram.

**UVLO detection equality**
If VIN decreases, R1, R2 value is expressed the following formula by the VINdet, the detect voltage of UVLO.

\[ R1 = R2[k\Omega] \times \frac{(VIN_{DET}[V] - 2.8[V])}{2.8[V]} \text{ [k}\Omega] \]

**UVLO release equality**
By using the R1, R2 in the above equality, the release voltage of UVLO can be expressed as following.

\[ VIN_{CAN} = 3.0[V] \times \frac{(R1[k\Omega] + R2[k\Omega])}{R2[k\Omega]} \text{ [V]} \]

[setting example]
If the normal input voltage, VIN is 24V, the detect voltage of UVLO is 18V, R2 is 30k ohm, R1 is calculated as following.

\[ R1 = R2[k\Omega] \times \frac{(VIN_{DET}[V] - 2.8[V])}{2.8[V]} = 30[k\Omega] \times \frac{(18[V] - 2.8[V])}{2.8[V]} = 163 \text{ [k}\Omega] \]

By using these R1, R2, the release voltage of UVLO, VINCan can be calculated as following.

\[ VIN_{CAN} = 3.0[V] \times \frac{(R1[k\Omega] + R2[k\Omega])}{R2[k\Omega]} = 3.0[V] \times \frac{30[k\Omega] + 163[k\Omega]}{30[k\Omega]} = 19.3 \text{ [V]} \]
3.4.6. how to set OVP, SCP

The OVP terminal is the input for over-voltage protection and short circuit protection of output voltage.

The OVP pin is high impedance, because the internal resistance to a certain bias is not connected.

So, the bias by the external components is required, even if OVP function is not used, because the open connection of this pin is not fixed the potential.

The resistor value can be calculated by the below formula, if the VOUT voltage is monitored, and that is divided by the resistor R1, R2 like the below diagram.

**OVP detection equality**
If the VOUT is boosted abnormally, VOVPdet is the detect voltage of OVP, R1, R2 can be expressed by the following formula.

\[
R1 = R2[k\Omega] \times \frac{\left(\text{VOVPdet}[V] - 3.0[V]\right)}{3.0[V]} \quad [k\Omega]
\]

**OVP release equality**
By using the R1, R2 in the above equality, the release voltage of OVP, VOVPcan can be expressed as following.

\[
\text{VOVPcan} = 2.9V \times \frac{\left(R1[k\Omega] + R2[k\Omega]\right)}{R2[k\Omega]} \quad [V]
\]

**SCP detection equality**
In the same way, the detect voltage of SCP, VSCPdet is

\[
\text{VSCPdet} = 0.2V \times \frac{\left(R1[k\Omega] + R2[k\Omega]\right)}{R2[k\Omega]} \quad [V]
\]

[setting example]
If the normal output voltage, VOUT is 40V, the detect voltage of OVP is 48V, R2 is 10k ohm, R1 is calculated as following.

\[
R1 = R2[k\Omega] \times \frac{\left(\text{VOVPdet}[V] - 3.0[V]\right)}{3.0[V]} = 10[k\Omega] \times \frac{\left(48[V] - 3.0[V]\right)}{3.0[V]} = 150 \quad [k\Omega]
\]

By using these R1, R2, the release voltage of OVP, VOVPcan can be calculated as following.

\[
\text{VOVPcan} = 2.9V \times \frac{\left(R1[k\Omega] + R2[k\Omega]\right)}{R2[k\Omega]} = 2.9V \times \frac{10[k\Omega] + 150[k\Omega]}{10[k\Omega]} = 46.4 \quad [V]
\]

Moreover, by using these R1, R2 the detect voltage of SCP, VSCPdet is

\[
\text{VSCPdet} = 0.2V \times \frac{\left(R1[k\Omega] + R2[k\Omega]\right)}{R2[k\Omega]} = 0.2V \times \frac{10[k\Omega] + 150[k\Omega]}{10[k\Omega]} = 3.2 \quad [V]
\]
3.4.7. how to set the interval until latch off

BD9285 built in the counter by latch off time, that is performed by counting the oscillation clock which is set by the RT pin. Since the common oscillation circuit is used for counting, the interval until latch off is corresponding to the 4096 clock, which the GATE pulse output continuously. Please refer the time chart of the operation from the detect abnormality to the latch off in the section 3.7.

Latch off time

BD9285 starts the counting up from the detection of each abnormal state, falls to the latch off state when the following interval has passed.

Only PWM=L input does not reset the timer counter, if the abnormal state continues.

\[
LATCH_{TIME} = 2^{12} \times \frac{R_{RT}[\Omega]}{1.5 \times 10^7} = 4096 \times \frac{R_{RT}[k\Omega]}{1.5 \times 10^7}[sec]
\]

Where \(LATCH_{TIME}\) is the interval until latch off state

\(R_{RT}\) is the connected resistor of RT pin.

[setting example]

If the resistor of RT pin is 100k ohm, the timer latch interval is as following.

\[
LATCH_{TIME} = 4096 \times \frac{R_{RT}[k\Omega]}{1.5 \times 10^7} = 4096 \times \frac{100[k\Omega]}{1.5 \times 10^7} = 27.3[msec]
\]
3.5. DCDC parts selection

3.5.1. how to set OCP / the calculation method for the current rating of DCDC parts

BD9285 stops the switching by the OCP detect, when the CS pin voltage is more than 0.5V. The resistor value of CS pin, \( R_{CS} \) need to be considered by the coil L current. And the current rating of DCDC external parts is required more than the peak current of the coil.

It is shown below that the calculation method of the coil peak current, the selection method of \( R_{CS} \) (the resistor value of CS pin) and the current rating of the external DCDC parts.

**(the calculation method of the coil peak current, \( I_{peak} \))**

At first, since the ripple voltage at CS pin depend on the application condition of DCDC, those put onto the equality to calculate as following.

The output voltage = \( V_{OUT} \) [V]

LED total current = \( I_{OUT} \) [A]

The DCDC input voltage of the power stage = \( V_{IN} \) [V]

The efficiency of DCDC = \( \eta \) [%]

And then, the averaged input current \( I_{IN} \) is calculated by the following equality

\[
I_{IN} = \frac{V_{OUT} [V] \times I_{OUT} [A]}{V_{IN} [V] \times \eta [%]} \quad [A]
\]

And the ripple current of the inductor L (\( \Delta I_L \) [A]) can be calculated by using DCDC the switching frequency, \( f_{SW} \), as following.

\[
\Delta I_L = \frac{(V_{OUT} [V] - V_{IN} [V]) \times V_{IN} [V]}{I[H] \times V_{OUT} [V] \times f_{SW} [Hz]} \quad [A]
\]

On the other hand, the peak current of the inductor \( I_{peak} \) can be expressed as the following equality.

\[
I_{peak} = I_{IN} [A] + \frac{\Delta I_L [A]}{2} \quad [A]
\]

Therefore, the bottom of the ripple current \( I_{min} \) is

\[
I_{min} = I_{IN} [A] - \frac{\Delta I_L [A]}{2}
\]

As \( I_{min} > 0 \), that operation mode is CCM (Continuous Current Mode), otherwise another mode is DCM (Discontinuous Current Mode).

**(the selection method of \( R_{CS} \))**

\( I_{peak} \) flows into \( R_{CS} \) and that cause the voltage signal to CS pin. (Please refer the right timing chart)

That peak voltage \( V_{CSpeak} \) is as following.

\[
V_{CSpeak} = R_{CS} \times I_{peak} \quad [V]
\]

As this \( V_{CSpeak} \) reaches to 0.5V, the DCDC output stops the switching.

Therefore, \( R_{CS} \) value is necessary to meet the under condition.

\[
R_{CS} \times I_{peak} [V] < 0.5[V]
\]

**(the current rating of the external DCDC parts)**

The peak current as the CS voltage reaches to OCP level (0.5V) is defined as \( I_{peak\_det} \).

\[
I_{peak\_det} = \frac{0.5[V]}{R_{CS} [\Omega]} \quad [A]
\]

The relation among \( I_{peak} \) (equality (1)), \( I_{peak\_det} \) (equality (2)) and the current rating of parts is required to meet the following

\[
I_{peak} << I_{peak\_det} \quad \text{The current rating of parts}
\]

Please make the selection of the external parts to meet the above condition such as FET, Inductor, diode.

[setting example]

The output voltage = \( V_{OUT} \) [V] = 40V

LED total current = \( I_{OUT} \) [A] = 0.48V
The DCDC input voltage of the power stage = VIN [V] = 24V
The efficiency of DCDC = \( \eta \) [%] = 90%
The averaged input current IIN is calculated as the following.

\[
I_{IN} [A] = \frac{V_{OUT} [V] \times I_{OUT} [A]}{V_{IN} [V] \times \eta \%} = \frac{40[V] \times 0.48[A]}{24[V] \times 90\%} = 0.89 \text{ [A]}
\]

And the ripple current of the inductor L (\( \Delta I_L[A] \)) can be calculated if the switching frequency, \( f_{SW} = 200\text{kHz} \), the inductor, \( L = 100\mu\text{H} \).

\[
\Delta I_L = \frac{(V_{OUT} [V] - V_{IN} [V]) \times V_{IN} [V]}{L[H] \times V_{OUT} [V] \times f_{SW} [Hz]} = \frac{(40[V] - 24[V]) \times 24[V]}{100 \times 10^{-6}[H] \times 40[V] \times 200 \times 10^{3}[Hz]} = 0.48 \text{ [A]}
\]

Therefore the inductor peak current, \( I_{peak} \) is

\[
I_{peak} = I_{IN} [A] + \frac{\Delta I_L [A]}{2} = 0.89[A] + \frac{0.48[A]}{2} = 1.13 \text{ [A]}
\]

The calculation result of the peak current

If \( Rcs \) is assume to be 0.3 ohm

\[
V_{CS_{peak}} = Rcs \times I_{peak} = 0.3[\Omega] \times 1.13[A] = 0.339 \text{ [V]} \ll 0.5V
\]

The above condition is met.

And \( I_{peak\_det} \), the current OCP works is

\[
I_{peak\_det} = \frac{0.5[V]}{0.3[\Omega]} = 1.67 \text{ [A]}
\]

If the current rating of the used parts is 2A,

\[
I_{peak} \ll I_{peak\_det} \ll \begin{align*}
\text{The current rating confirmation of DCDC parts}
\end{align*}
\]

This inequality meets the above relationship. The parts selection is proper.

And \( I_{min} \), the bottom of the \( I_L \) ripple current can be calculated as following.

\[
I_{MIN} = I_{IN} [A] - \frac{\Delta I_L [A]}{2} = 1.13[A] - 0.48[A] = 0.65[A] \gg 0
\]

This inequality implies the operation is the continuous current mode.
3.5.2. Inductor selection

The inductor value affects the input ripple current. The equality in the section 3.5.1 is as following.

\[ \Delta I_L = \frac{(V_{OUT}[V] - V_{IN}[V]) \times V_{IN}[V]}{L[H] \times f_{SW}[Hz]} \] [A]

\[ I_{IN} = \frac{V_{OUT}[V] \times I_{OUT}[A]}{V_{IN}[V] \times \eta[\%]} \] [A]

\[ I_{peak} = I_{IN}[A] + \frac{\Delta I_L[A]}{2} \] [A]

Where
- L: the coil inductance [H]
- Vout: the DCDC output voltage [V]
- Vin: the input voltage [V]
- Iout: the output load current (the summation of LED current) [A]
- Iin: the input current [A]
- Fsw: the oscillation frequency [Hz]

If in the continuous current mode, Please set \( \Delta I_L \) to 30% - 50% of the output load current.

* The current exceeding the rated current value of inductor flown through the coil causes magnetic saturation, results in decreasing in efficiency. Inductor needs to be selected to have such adequate margin that peak current does not exceed the rated current value of the inductor.
* To reduce inductor loss and improve efficiency, inductor with low resistance components (DCR, ACR) needs to be selected

3.5.3. Output capacitance Cout selection

Output capacitor needs to be selected in consideration of equivalent series resistance required to even the stable area of output voltage or ripple voltage. Be aware that set LED current may not be flown due to decrease in LED terminal voltage if output ripple component is high. Output ripple voltage \( \Delta V_{OUT} \) is determined by Equation (4):

\[ \Delta V_{OUT} = I_{LMAX} \times R_{ESR} + \frac{1}{C_{OUT}} \times I_{OUT} \times \frac{1}{\eta} \times f_{SW} [V] \] (4)

where, \( R_{ESR} \) is the equivalent series resistance of Cout.

* Rating of capacitor needs to be selected to have adequate margin against output voltage.
* To use an electrolytic capacitor, adequate margin against allowable current is also necessary. Be aware that the LED current is larger than the set value transitionally in case that LED is provided with PWM dimming especially.

3.5.4. MOSFET selection

Though there is no problem if the absolute maximum rating is larger than the rated current of the inductor \( L \), or is larger than the sum of the tolerance voltage of \( C_{OUT} \) and the rectifying diode \( V_f \). The product with small gate capacitance (injected charge) needs to be selected to achieve high-speed switching.

* One with over current protection setting or higher is recommended.
* The selection of one with small on resistance results in high efficiency.

3.5.5. Rectifying diode selection

A schottky barrier diode which has current ability higher than the rated current of \( L \), the reverse voltage larger than the tolerance voltage of \( C_{OUT} \), and the low forward voltage \( V_f \) especially needs to be selected.
3.6. Loop compensation

A current mode DCDC converter has each one pole (phase lag) \( f_p \) due to CR filter composed of the output capacitor and the output resistance (= LED current) and zero (phase lead) \( f_Z \) by the output capacitor and the ESR of the capacitor.

Moreover, a step-up DCDC converter has RHP zero (right-half plane zero point) \( f_{\text{RHP}} \) which is unique with the boost converter. This zero may cause the unstable feedback. To avoid this by RHP zero, the loop compensation that the cross-over frequency \( f_c \) set as following, is suggested.

\[
f_c = \frac{f_{\text{RHP}}}{5} \quad (f_{\text{RHP}}: \text{RHP zero frequency})
\]

Considering the response speed, the below calculated constant is not always optimized completely. It needs to be adequately verified with an actual device.

\[
f_p = \frac{I_{\text{LED}}}{2\pi \times V_{\text{OUT}} \times C_{\text{OUT}}} \quad \text{[Hz]}
\]

\[
f_{\text{RHP}} = \frac{V_{\text{OUT}} \times (1 - D)^2}{2\pi \times L \times I_{\text{LED}}} \quad \text{[Hz]}
\]

Where \( I_{\text{LED}} \) is the summation of LED current. (Continuous Current Mode)

\[
D = \frac{V_{\text{OUT}} - V_{\text{IN}}}{V_{\text{OUT}}}
\]

Where \( g_m = 4.0 \times 10^{-4} \text{[S]} \)

\[
R_{\text{FB1}} = \frac{f_{\text{RHP}} \times R_{\text{CS}} \times I_{\text{LED}}}{5 \times f_p \times g_m \times V_{\text{OUT}} \times (1 - D)} \quad \text{[\Omega]}
\]

\[
C_{\text{FB1}} = \frac{1}{2 \pi \times R_{\text{FB1}} \times f_p} \quad \text{[F]}
\]

\[
C_{\text{FB2}} = \frac{R_{\text{ESR}} \times C_{\text{OUT}}}{R_{\text{FB1}}} \quad \text{[F]}
\]

*When a ceramic capacitor (with \( R_{\text{ESR}} \) of the order of milliohm) is used to \( C_{\text{OUT}} \), the operation is stabilized by insertion of \( C_{\text{FB2}} \).

To improve the transient response, \( R_{\text{FB1}} \) need to be increase, \( C_{\text{FB1}} \) need to be decrease. It needs to be adequately verified with an actual device in consideration of vary from parts to parts since phase margin is decreased.
### 3.7. Timing chart

#### 3.7.1 starting up 1 (STB inputs and PWM signal succeed)

- **VCC**: 7.5V
- **STB**: 2.5V
- **PWM1 and PWM2**: 2.4V
- **SS**: 3.0V
- **GATE**: 2.5V
- **FAILB**: 1.1V

(*1) TC54 starts up if STB turns from L to H. The pin SS is not charged in the state that the PWM signal is not input, the boost is not started.

(*2) The charge of the pin SS starts by the positive edge of PWM=L to H, and the soft start starts. The GATE pulse outputs only during PWM=H. And as the SS is less than 1.1V, the pulse does not output. The pin SS continues charging in spite of the assertion of PWM and OVP.

(*3) The soft start interval will end if the voltage of the pin SS, Vss reaches to 3.0V. By this time, BD9285 boost Vout where the set LED current flows. It is started to monitor the abnormal detection of SCP and OPEN.

(*4) As STB=L, instantaneously the boost operation is stopped. (GATE=L, SS=L) On the other hand, the discharge circuit works in the interval "STB=L and V5UVLO=H". Please refer to the time chart in the section 3.7.3 for details.

(*5) As STB=H again, the boost operation restarts by the next PWM=L to H. It is the same operation as the timing of (*1). Please refer to the section 3.4.1 for the setting of soft start external capacitance.
3.7.2 starting up 2 (PWM signal inputs and STB succeeds)

(*1)...TC54 starts up if STB turns from L to H.

(*2)...At the moment the release of V5UVLO (the UVLO of the pin TC54), or the time of the positive edge of PWM=L to H, the soft start starts. The GATE pulse outputs only during PWM1=H. And as the SS is less than 1.1V, the pulse does not output. The pin SS continues charging in spite of the assertion of PWM and OVP.

(*3)...The soft start interval will end if the voltage of the pin SS, VSS reaches to 3.0V. By this time, BD9285 boost Vout where the set LED current flows. It is started to monitor the abnormal detection of SCP and OPEN.

(*4)...As STB=L, instantaneously the boost operation is stopped. (GATE=L, SS=L) On the other hand, the discharge circuit works in the interval "STB=L and V5UVLO=H". Please refer to the time chart in the section 3.7.3 for details.

(*5)...As STB=H again, it is the same operation as the timing of (*1).
3.7.3 turn off

(1) As STB pin turns High to Low, BD9285F stops the boost operation, starts the discharge of TC54.
(2) During STB=L and V5UVLO=H, the DIMOUT asserts the same logic of PWM. TC54=5.4V is discharged until 2.4V by the constant current 10uA. And IC turns off. Vout need to be discharged adequately so that LED does not turns on drastically at the next start up.

For detailed instructions, please refer the section 3.4.2 “how to shutdown and set TC54 capacitance”
3.7.4 the soft start function

(*1)...The SS pin charge does not start by just STB=H. "PWM1=H and PWM2=H" is required to start the soft start. In the low SS voltage, the GATE pin duty is depend on the SS voltage. And as the SS is less than 1.1V, the pulse does not output.

(*2)...By the low STB=L, the SS pin is discharged immediately.

(*3)...As the STB recovered to STB=H, The SS charge starts immediately by the logic “PWM1 and PWM2=H” in this chart.

(*4)...The SS pin is discharged immediately by the UVLO=L.

(*5)...The SS pin is discharged immediately by the VCCUVLO=L

(*6)...The SS pin is discharged immediately by the TC54UVLO=L

(*7)...The SS pin is not discharged by the abnormal detection of the latch off type such as OVP until the latch off...
3.7.5 the OVP detection

(*1)...As OVP is detected, the output GATE=L, DIMOUT=L, and the CP counter starts.

(*2)...If OVP is released within 4 clock of CP counter of the GATE pin frequency, the boost operation restarts.

(*3)...As the OVP is detected again, the boost operation is stopped.

(*4)...As the OVP detection continues up to 4 count by the CP counter, IC will be latched off.

(*5)...As the latched off, the boost operation doesn't restart even if OVP is released.

(*6)...The STB=L input can make IC reset. In this chart, DIMOUT asserts high by the discharge function in the paragraph 3.7.3.

(*7)...It normally starts as STB turns L to H.

(*8)...The operation of the OVP detection is not related to the logic of PWM.
3.7.6 the SCP detection

(*1) ...During the soft start, the detection of SCP is masked.
(*2) ...As SCP is detected, the output GATE=L, DIMOUT=L, and the CP counter starts.
(*3) ...If SCP is released within 4096 clock of CP counter of the GATE pin frequency, the boost operation restarts.
(*4) ...As the SCP is detected again, the boost operation is stopped.
(*5) ...As the SCP detection continues up to 4096 count by the CP counter, IC will be latched off.
(*6) ...The STB=L input can make IC reset. In this chart, DIMOUT asserts high by the discharge function in the paragraph 3.7.3.
(*7) ...It normally starts as STB turns L to H.
(*8) ...The operation of the SCP detection is not related to the logic of PWM.
3.7.7 LED OPEN detection

(*1)...During starting up, even if the normality, ISENSE<0.1V because of the low Vout. Therefore the OPEN detection will be masked for the soft start period.
(*2)...In the same way, as PWM=L, ISENSE<0.1V because of DIMOUT=L. OPEN will be masked, too.
(*3)...Though the OPEN is detected if ISENSE<0.1V as the PWM=H, it is not judged immediately to abnormal state. The behavior of GATE, FAILB keeps the normal operation.
(*4)...The CP counter will start if the OPEN detection continues 4 clock of the GATE frequency. To detect the OPEN state continuously, it compulsorily becomes DIMOUT=H regardless of the PWM logic.
(*5)...When the OPEN detection continues up to 4649 count with the CP counter, IC will be latched off. At this time, it asserts GATE=L, DIMOUT=L, FAILB=L for the first time.
(*6)...The latch off state can be reset by the STB=L.
(*7)...It normally starts by STB=L to H, in this figure.
3.7.8 LED OCP detection

(*1)...If ISENSE > 4.0V, LEDOCP is detected, it becomes GATE=L. To detect LEDOCP continuously, the DIMOUT is compulsorily high, regardless of the PWM dimming signal.

(*2)...When the LEDOCP releases within the GATE frequency 4096 counts of the CP counter, the boost operation restarts.

(*3)...As the LEDOCP is detected again, the boost operation is stopped, too.

(*4)...If the LEDOCP detection continues up to 4096 count with the CP counter, IC will be latched off.

(*5)...Once IC is latched off, the boost operation doesn't restart even if the LEDOCP releases.

(*6)...The latch off state can be reset by the STB=L. In this chart, DIMOUT asserts high by the discharge function in the paragraph 3.7.3.

(*7)...It normally starts by STB=L to H.

(*8)...The operation of the LEDOCP detection is not related to the logic of the PWM.
3.7.9 the spontaneous detection OVP and OPEN.

(*1)...The time chart shows the OPEN detects faster and does not reach to the latch off state. The DIMOUT asserts high.
(*2)...If OPEN and OVP is detected spontaneously, OVP has the priority, and GATE=L, DIMOUT=L.
(*3)...IC will be latched off by the OVP factor.
(*4)...The latch off state is reset by the STB=L.
(*5)...The OVP has the priority too, in the case the OVP is detected first and the OPEN succeeds.
Operational Notes

1.) This product is produced with strict quality control, but might be destroyed if used beyond its absolute maximum ratings including the range of applied voltage or operation temperature. Failure status such as short-circuit mode or open mode can not be estimated. If a special mode beyond the absolute maximum ratings is estimated, physical safety countermeasures like fuse needs to be provided.

2.) Connecting the power line to IC in reverse polarity (from that recommended) may cause damage to IC. For protection against damage caused by connection in reverse polarity, countermeasures, installation of a diode between external power source and IC power terminal, for example, needs to be taken.

3.) When this product is installed on a printed circuit board, attention needs to be paid to the orientation and position of IC. Wrong installation may cause damage to IC. Short circuit caused by problems like foreign particles entering between outputs or between an output and power GND also may cause damage.

4.) Since the back electromotive force of external coil causes regenerated current to return, countermeasures like installation of a capacitor between power source and GND as the path for regenerated current needs to be taken. The capacitance value must be determined after it is adequately verified that there is no problem in properties such that the capacity of electrolytic capacitor goes down at low temperatures. Thermal design needs to allow adequate margin in consideration of allowable loss (Pd) in actual operation state.

5.) The GND pin needs to be at the lowest potential in any operation state.

6.) Thermal design needs to be done with adequate margin in consideration of allowable loss (Pd) in actual operation state.

7.) Use in a strong magnetic field may cause malfunction.

8.) Output Tr needs to not exceed the absolute maximum rating and ASO while using this IC. As CMOS IC and IC which has several power sources may undergo instant flow of rush current at turn-on, attention needs to be paid to the capacitance of power source coupling, power source, and the width and run length of GND wire pattern.

9.) This IC includes temperature protection circuit (TSD circuit). Temperature protection circuit (TSD circuit) strictly aims blockage of IC from thermal runaway, not protection or assurance of IC. Therefore use assuming continuous use and operation after this circuit is worked needs to not be done.

10.) As connection of a capacitor with a pin with low impedance at inspection of a set board may cause stress to IC, discharge needs to be performed every one process. Before a jig is connected to check a process, the power needs to be turned off absolutely. Before the jig is removed, as well, the power needs to be turned off.

11.) This IC is a monolithic IC which has P+ isolation for separation of elements and P board between elements. A P-N junction is formed in this P layer and N layer of elements, comprising various parasitic elements. For example, a resistance and transistor are connected to a terminal as shown in the figure.

○ When GND>(Terminal A) in the resistance and when GND>(Terminal B) in the transistor (NPN), P-N junction operates as a parasitic diode.
○ When GND>(Terminal B) in the transistor (NPN), parasitic NPN transistor operates in N layer of other elements nearby the parasitic diode described before.

Parasitic elements are formed by the structure of IC. Operation of parasitic elements can cause mutual interference among circuits, malfunction as well as damage. Therefore such use as will cause operation of parasitic elements like application of voltage on the input terminal lower than GND (P board) need to not be done.

Status of this document
The Japanese version of this document is formal specification. A customer may use this translation version only for a reference to help reading the formal version.
If there are any differences in translation version of this document formal version takes priority.

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TSZ22111 - 15 - 001
35/36
TSZ02201-0F1F0C10030-1-2
11.Jul.2012 Rev.001
Not Recommended for New Designs
**Ordering Information**

<table>
<thead>
<tr>
<th>Part Number</th>
<th>Package</th>
</tr>
</thead>
<tbody>
<tr>
<td>BD9285F</td>
<td>F:SOP18</td>
</tr>
</tbody>
</table>

Packaging and forming specification
XX: Please confirm the formal name to our sales.

**Physical Dimension Tape and Reel Information**

**SOP18**

### Marking Diagram (TOP VIEW)

**SOP18(TOP VIEW)**

- **Part Number Marking**
- **LOT Number**
- **1PIN MARK**

**<Tape and Reel information>**

<table>
<thead>
<tr>
<th>Tape</th>
<th>Embossed carrier tape</th>
</tr>
</thead>
<tbody>
<tr>
<td>Quantity</td>
<td>2000pcs</td>
</tr>
<tr>
<td>Direction of feed</td>
<td>E2</td>
</tr>
</tbody>
</table>

The direction is the 1pin of product is at the upper left when you hold reel on the left hand and you pull out the tape on the right hand.

Order quantity needs to be multiple of the minimum quantity.

**Not Recommended for New Designs**
Notice

Precaution on using ROHM Products

1. Our Products are designed and manufactured for application in ordinary electronic equipments (such as AV equipment, OA equipment, telecommunication equipment, home electronic appliances, amusement equipment, etc.). If you intend to use our Products in devices requiring extremely high reliability (such as medical equipment (Note 1), transport equipment, traffic equipment, aircraft/spacecraft, nuclear power controllers, fuel controllers, car equipment including car accessories, safety devices, etc.) and whose malfunction or failure may cause loss of human life, bodily injury or serious damage to property ("Specific Applications"), please consult with the ROHM sales representative in advance. Unless otherwise agreed in writing by ROHM in advance, ROHM shall not be in any way responsible or liable for any damages, expenses or losses incurred by you or third parties arising from the use of any ROHM’s Products for Specific Applications.

(Note 1) Medical Equipment Classification of the Specific Applications

<table>
<thead>
<tr>
<th>JAPAN</th>
<th>USA</th>
<th>EU</th>
<th>CHINA</th>
</tr>
</thead>
<tbody>
<tr>
<td>CLASS III</td>
<td>CLASS III</td>
<td>CLASS II b</td>
<td>CLASS III</td>
</tr>
</tbody>
</table>

2. ROHM designs and manufactures its Products subject to strict quality control system. However, semiconductor products can fail or malfunction at a certain rate. Please be sure to implement, at your own responsibilities, adequate safety measures including but not limited to fail-safe design against the physical injury, damage to any property, which a failure or malfunction of our Products may cause. The following are examples of safety measures:

[a] Installation of protection circuits or other protective devices to improve system safety
[b] Installation of redundant circuits to reduce the impact of single or multiple circuit failure

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[a] Use of our Products in any types of liquid, including water, oils, chemicals, and organic solvents
[b] Use of our Products outdoors or in places where the Products are exposed to direct sunlight or dust
[c] Use of our Products in places where the Products are exposed to sea wind or corrosive gases, including Cl₂, H₂S, NH₃, SO₂, and NO₂
[d] Use of our Products in places where the Products are exposed to static electricity or electromagnetic waves
[e] Use of our Products in proximity to heat-producing components, plastic cords, or other flammable items
[f] Sealing or coating our Products with resin or other coating materials
[g] Use of our Products without cleaning residue of flux (even if you use no-clean type fluxes, cleaning residue of flux is recommended), or Washing our Products by using water or water-soluble cleaning agents for cleaning residue after soldering
[h] Use of the Products in places subject to dew condensation

4. The Products are not subject to radiation-proof design.

5. Please verify and confirm characteristics of the final or mounted products in using the Products.

6. In particular, if a transient load (a large amount of load applied in a short period of time, such as pulse. is applied, confirmation of performance characteristics after on-board mounting is strongly recommended. Avoid applying power exceeding normal rated power, exceeding the power rating under steady-state loading condition may negatively affect product performance and reliability.

7. De-rate Power Dissipation (Pd) depending on Ambient temperature (Ta). When used in sealed area, confirm the actual ambient temperature.

8. Confirm that operation temperature is within the specified range described in the product specification.

9. ROHM shall not be in any way responsible or liable for failure induced under deviant condition from what is defined in this document.

Precaution for Mounting / Circuit board design

1. When a highly active halogenous (chlorine, bromine, etc.) flux is used, the residue of flux may negatively affect product performance and reliability.

2. In principle, the reflow soldering method must be used; if flow soldering method is preferred, please consult with the ROHM representative in advance.

For details, please refer to ROHM Mounting specification
Precautions Regarding Application Examples and External Circuits
1. If change is made to the constant of an external circuit, please allow a sufficient margin considering variations of the characteristics of the Products and external components, including transient characteristics, as well as static characteristics.

2. You agree that application notes, reference designs, and associated data and information contained in this document are presented only as guidance for Products use. Therefore, in case you use such information, you are solely responsible for it and you must exercise your own independent verification and judgment in the use of such information contained in this document. ROHM shall not be in any way responsible or liable for any damages, expenses or losses incurred by you or third parties arising from the use of such information.

Precaution for Electrostatic
This Product is electrostatic sensitive product, which may be damaged due to electrostatic discharge. Please take proper caution in your manufacturing process and storage so that voltage exceeding the Products maximum rating will not be applied to Products. Please take special care under dry condition (e.g. Grounding of human body / equipment / solder iron, isolation from charged objects, setting of Ionizer, friction prevention and temperature / humidity control).

Precaution for Storage / Transportation
1. Product performance and soldered connections may deteriorate if the Products are stored in the places where:
   [a] the Products are exposed to sea winds or corrosive gases, including Cl2, H2S, NH3, SO2, and NO2
   [b] the temperature or humidity exceeds those recommended by ROHM
   [c] the Products are exposed to direct sunshine or condensation
   [d] the Products are exposed to high Electrostatic

2. Even under ROHM recommended storage condition, solderability of products out of recommended storage time period may be degraded. It is strongly recommended to confirm solderability before using Products of which storage time is exceeding the recommended storage time period.

3. Store / transport cartons in the correct direction, which is indicated on a carton with a symbol. Otherwise bent leads may occur due to excessive stress applied when dropping of a carton.

4. Use Products within the specified time after opening a humidity barrier bag. Baking is required before using Products of which storage time is exceeding the recommended storage time period.

Precaution for Product Label
QR code printed on ROHM Products label is for ROHM’s internal use only.

Precaution for Disposition
When disposing Products please dispose them properly using an authorized industry waste company.

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Since our Products might fall under controlled goods prescribed by the applicable foreign exchange and foreign trade act, please consult with ROHM representative in case of export.

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